

FHDW01C

High Precision Protection Circuit for Li-ion or Li-polymer Battery

**FEATURES**

- Ideal protection circuit for one-cell Li-ion or Li-polymer battery
- High precision protection voltage threshold (over-charge/over-discharge)
- Allow or inhibit low power consumption mode
- High precision over-discharge protection current threshold
- Protection for battery short
- Multi-type of detector voltage and time delay option
- Allow or inhibit variable 0V battery charge
- Very few external elements
- Small SOT23-6 Package

**APPLICATIONS**

- Protection circuit for charge and discharge of Li-ion or Li-polymer battery
- High precision protector for cell phone battery and any other protector of Li-ion or Li-polymer battery

**DESCRIPTION**

FHDW01C series are high precision protection ICs for over-charge and over-discharge of rechargeable one-cell Li-ion or Li-polymer battery. It integrates the high precision protection capability for over-charge, over-discharge, excess-current discharge, and battery short.

Under normal conditions, when  $V_{DD}$  is between the protection thresholds of over-charge ( $V_{OC}$ ) and over-discharge ( $V_{OD}$ ), and the detection voltage of  $V_M$  is between the charger detect voltage ( $V_{CHG}$ ) and excess-current discharge ( $V_{EDI}$ ), the outputs of  $C_{OUT}$  and  $D_{OUT}$  are high conducting the N-MOSFET charge controller, Q1, and the N-MOSFET discharge controller, Q2. Thus, the battery can be charged through a charger and can be discharged through a load.

FHDW01C series realizes the over-charge and over-discharge protection through detecting the voltages of  $V_{DD}$  and  $V_M$ . When abnormal conditions occur during charging or discharging, the outputs of  $C_{OUT}$  and  $D_{OUT}$  both change from a high level to a low level, stopping charging or discharging by turning Q1/Q2 off.

All protections can be released at corresponding conditions. When the recovery condition is met, the outputs of  $C_{OUT}/D_{OUT}$  change from a low level to a high level, turning on Q1/Q2 to enable charge/discharge.

FHDW01C sets internal delay time for each protection and release. It does not enter into the protection or release state until its corresponding condition reaches its delay time. If the protection or release condition disappears in less than the corresponding delay time, it will not enter to either the protection or release state.

**PIN CONFIGURATIONS**

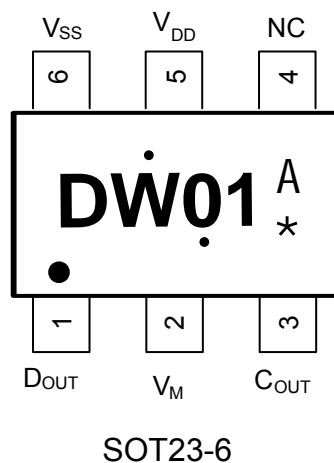


Figure 1 FHDW01C Pin Configurations (Not to scale, \* & • there are changes)

## ORDERING INFORMATION

[Table 2] Detector Voltage Threshold and Delay Time

PARAMETER NAME	VALUE	ACCURACY RANGE
Protection threshold of over-charge $V_{OCTYP}$	4.300V	±25mV
Release threshold of over-charge $V_{OVRTYP}$	4.100V	±50mV
Protection threshold of over-discharge $V_{ODTYP}$	2.500V	±75mV
Release threshold of over-discharge $V_{ODRTYP}$	2.900V	±75mV
Protection threshold of excess-current discharge $V_{EDITYP}$	0.170V	±20mV
Protection delay time of over-charge $t_{OCTYP}$	110ms	±30%
Protection delay time of over-discharge $t_{ODTYP}$	55ms	±30%
Protection delay time of excess-current discharge $t_{EDITYP}$	7.0ms	±30%
0V-charge	Yes	
Low power consumption mode	Yes	

## FUNCTIONAL DIAGRAM

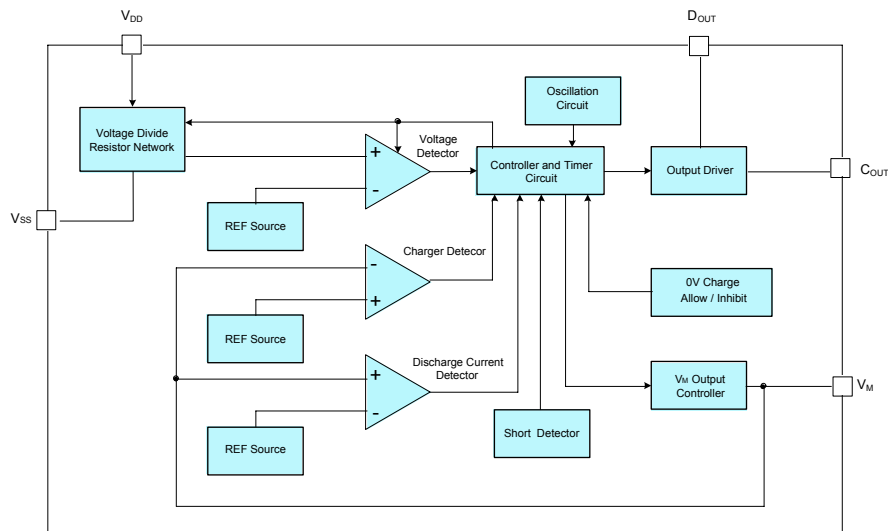


Figure 2 FHDW01C Functional Diagram

## PIN DESCRIPTION

[Table 3] PIN Description

NAME	ORDER	I/O	FUNCTION
D <sub>OUT</sub>	1	O	<b>Discharge Control Output</b> Connect to the Gate of the external discharge controller N-MOSFET Q2.
V <sub>M</sub>	2	I	<b>Charge/Discharge Current Sense Input</b> Connect this to the Source of external charge controller N-MOSFET Q1 by a resistance (normally 1kΩ), then the voltage drop on Q1 and Q2, which cause by the charge/discharge current can be sensed.
C <sub>OUT</sub>	3	O	<b>Charge Control Output</b> Connect to the Gate of the external charge controller N-MOSFET Q1.
NC	4		<b>Not Connected</b>
V <sub>DD</sub>	5	POW	<b>Power Supply Input</b> Connect to the positive of power supply (battery normally), a 0.1μF ceramic capacitor is required for decoupling.
V <sub>SS</sub>	6	POW	<b>Ground</b> Connect to the negative of power supply.

## ABSOLUTE MAXIMUM RATINGS

Power supply V <sub>DD</sub> .....	-0.3V~+10V	Storage temperature .....	-65°C~150°C
V <sub>M</sub> ,C <sub>OUT</sub> acceptable voltage..	V <sub>DD</sub> -35V~V <sub>DD</sub> +0.3V	Power consumption P <sub>D</sub> (T <sub>A</sub> =25°C)	
D <sub>OUT</sub> acceptable voltage .....	-0.3V~V <sub>DD</sub> +0.3V	SOT23-6 package (θ <sub>JA</sub> =200°C/W).....	625mW
Operation temperature T <sub>A</sub> .....	-40°C~+85°C	Solder Temperature (Tin soldering, 10s).....	260°C
Junction temperature .....	150°C		



**Note:** Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond the recommended operating condition are not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

## ELECTRICAL SPECIFICATION

( $V_{DD} = 3.6V$ ,  $T_A = 25^\circ C$ , unless otherwise specified. The operation temperature with Mark “♦” is:  $-40^\circ C \leq T_A \leq 85^\circ C$ )

[Table 4] Electrical Specification

PARAMETER	SYMBOL	CONDITION		MIN	TYP	MAX	UNIT
Power supply	$V_{DD}$		♦	1.5		10	V
Over-charge protection threshold (rising)	$V_{OC}$			$V_{OC\text{TYP}}-0.025$	$V_{OC\text{TYP}}$	$V_{OC\text{TYP}}+0.025$	V
			♦	$V_{OC\text{TYP}}-0.080$	$V_{OC\text{TYP}}$	$V_{OC\text{TYP}}+0.080$	V
Over-charge release Threshold(falling)	$V_{OCR}$	R1=100Ω (note)		$V_{OCR\text{TYP}}-0.050$	$V_{OCR\text{TYP}}$	$V_{OCR\text{TYP}}+0.050$	V
			♦	$V_{OCR\text{TYP}}-0.080$	$V_{OCR\text{TYP}}$	$V_{OCR\text{TYP}}+0.080$	V
Over-charge protection delay time	$t_{OC}$	$V_{DD}=3.6V \rightarrow 4.4V$		$0.7 \times t_{OC\text{TYP}}$	$t_{OC\text{TYP}}$	$1.3 \times t_{OC\text{TYP}}$	ms
Over-discharge protection Threshold(falling)	$V_{OD}$			$V_{OD\text{TYP}}-0.075$	$V_{OD\text{TYP}}$	$V_{OD\text{TYP}}+0.075$	V
			♦	$V_{OD\text{TYP}}-0.105$	$V_{OD\text{TYP}}$	$V_{OD\text{TYP}}+0.105$	V
Over-discharge release Threshold(rising)	$V_{ODR}$			$V_{ODR\text{TYP}}-0.075$	$V_{ODR\text{TYP}}$	$V_{ODR\text{TYP}}+0.075$	V
			♦	$V_{ODR\text{TYP}}-0.105$	$V_{ODR\text{TYP}}$	$V_{ODR\text{TYP}}+0.105$	V
Over-discharge protection delay time	$t_{OD}$	$V_{DD}=3.6V \rightarrow 2.4V$		$0.7 \times t_{OD\text{TYP}}$	$t_{OD\text{TYP}}$	$1.3 \times t_{OD\text{TYP}}$	ms
Excess-current discharge protection threshold	$V_{EDI}$			$V_{EDI\text{TYP}}-0.020$	$V_{EDI\text{TYP}}$	$V_{EDI\text{TYP}}+0.020$	V
Excess-current discharge protection delay time	$t_{EDI}$			$0.7 \times t_{EDI\text{TYP}}$	$t_{EDI\text{TYP}}$	$1.3 \times t_{EDI\text{TYP}}$	ms
Excess-current discharge release delay time	$t_{EDIR}$			1.20	1.80	2.40	ms
Battery short protection threshold	$V_{SHORT}$	Voltage of $V_M$		0.82	1.36	1.75	V
Battery short protection delay time	$t_{SHORT}$			200	400	600	μs
Charger detect voltage	$V_{CHG}$	$V_{DD}=3.0V$		-0.27	-0.5	-0.86	V
Resistance of $V_M$ to $V_{DD}$	$R_{VMD}$	$V_{DD}=1.8V$ , $V_M=0V$		100	300	900	kΩ
Resistance of $V_M$ to $V_{SS}$	$R_{VMS}$			15	30	45	kΩ
$C_{OUT}$ output low level pull-low resistor					4		MΩ
$C_{OUT}$ output high level		$V_{DD}=3.9V$ , $I_{COUT}=10\mu A$		$V_{DD}-0.4$	$V_{DD}-0.2$		V
$D_{OUT}$ output low level		$V_{DD}=2.0V$ , $I_{DOUT}=10\mu A$			0.2	0.4	V
$D_{OUT}$ output high level		$V_{DD}=3.9V$ , $I_{DOUT}=10\mu A$		$V_{DD}-0.4$	$V_{DD}-0.2$		V
Power current	$I_{DD}$	$V_{DD} = 3.9V$			2.0	6.0	μA
Current under low power consumption mode	$I_{PDWN}$	$V_{DD}=2.0V$			0.7	1.0	μA
0V charge allow threshold (If 0V charge allow)	$V_{0V\_CHG}$	Charger Voltage		1.2			V
0V charge inhibit threshold (If 0V charge inhibit)	$V_{0V\_INH}$	Battery Voltage, $V_M=-2.0V$				1.2	V



**Note:** 1. All the voltages are referred to  $V_{SS}$ , unless otherwise specified.  
2. Shown in Figure 3.

## FUNCTION DESCRIPTION

FHDW01C is a high precision protection circuit for the one-cell Li-ion or Li-polymer battery. Under normal conditions, during the battery charging, FHDW01C may get into the over-charge protection. It resets to the normal condition when it reaches the release condition. During the battery discharging, FHDW01C may get into the over-discharge or excess-current discharge protection. It can also reset to the normal state when it reaches the release condition. Figure 3 shows the typical application schematic. The state conversion diagram is shown in Figure 4. The detailed description of each condition is followed.

### Normal Condition

Under normal conditions, FHDW01C is powered by the battery. When  $V_{DD}$  is between the protection thresholds of over-charge ( $V_{OC}$ ) and over-discharge ( $V_{OD}$ ),  $V_M$  is between the charger detect voltage ( $V_{CHG}$ ) and excess-current discharge ( $V_{EDI}$ ), the outputs of  $C_{OUT}$  and  $D_{OUT}$  become high and turn on the charge controller N-MOSFET Q1 and the discharge controller N-MOSFET Q2. Thus, the battery can be charged through a charger or discharged through a load.

### Over-charge Protection

#### • Protection condition

During the battery charging and under the normal condition, if the voltage of  $V_{DD}$  exceeds the over-charge protection threshold ( $V_{OC}$ ) and this state lasts more than the over-charge protection delay time ( $t_{OC}$ ), the voltage of  $C_{OUT}$  pin is equal to the voltage of  $V_M$  pin. The N-MOSFET's charge controller Q1 is turned off. The charge current is "shut off". FHDW01C gets into over-charge protection.

#### • Release condition

FHDW01C can recover from over-charge protection when it meets one of the following two conditions. 1) The battery discharges itself to make  $V_{DD}$  lower than the over-charge release threshold ( $V_{OCR}$ ); 2) The battery is discharged through an extra load (Note: Even though Q1 is turned off, discharge loop is still available due to its body diode),  $V_{DD}$  is lower than the over-charge protection threshold ( $V_{OC}$ ), and the voltage of  $V_M$  pin is higher than the excess-current discharge protection threshold ( $V_{EDI}$ ). (Before Q1 is turned on, the voltage of  $V_M$  is one diode voltage higher than the voltage of  $V_{SS}$ ).

After FHDW01C recovers to normal condition, the output of  $C_{OUT}$  pin goes to a high level. The charge controller N-MOSFET, Q1, is turned on again.

Once FHDW01C enter into over-charge protection, it will never release to normal condition if a charger is always connected, even if its  $V_{DD}$  is below  $V_{ODR}$ . It only can be released by disconnecting the charger.

### Over-discharge Protection/Low Power Consumption Mode

#### • Protection condition

Under normal conditions, if the voltage of  $V_{DD}$  pin is

lower than the over-discharge protection threshold ( $V_{OD}$ ) and this state lasts more than the over-discharge protection delay time ( $t_{OD}$ ), the voltage of  $D_{OUT}$  pin goes to low ( $V_{SS}$ ) from a high level. The discharge controller N-MOSFET, Q2, is turned off, shutting off the discharging loop. FHDW01C gets into the over-discharge protection. The voltage of  $V_M$  pin is pulled up to  $V_{DD}$  through the internal resistor,  $R_{VMD}$ .

During over-charge protection, the voltage of  $V_M$  pin (equal to  $V_{DD}$ ) is always higher than the battery short protection threshold ( $V_{SHORT}$ ). Thus, the circuit gets into a low power consumption or "Power saving" mode. In this mode, the current of  $V_{DD}$  pin is less than  $0.7\mu A$ .

#### • Release condition

In the low power consumption mode, the battery should be charged to make the voltage of  $V_M$  pin lower than the battery short protection threshold ( $V_{SHORT}$ ), and then FHDW01C can recover to the over-voltage discharge protection. (The charging circuit is still available due to the diode in Q2). Under this condition, the output level of  $D_{OUT}$  is held low, and Q2 is still turned off. If stopped from charging, FHDW01C returns to the low power consumption mode, because the voltage of  $V_M$  pin is still pulled up to  $V_{DD}$  by the  $R_{VMD}$  resistor and the voltage is higher than the battery short protection threshold ( $V_{SHORT}$ ). Only when the battery is charged continually until the voltage of  $V_{DD}$  pin rises above the over-discharge protection threshold ( $V_{OD}$ ), FHDW01C can recover to the normal condition from the over-discharge protection.

FHDW01C also can release to the normal condition from the over-discharge protection, if the battery's self-voltage lifting feature makes the voltage of  $V_{DD}$  higher than the over-discharge release threshold ( $V_{ODR}$ ). After FHDW01C recovers to the normal condition, the output of  $D_{OUT}$  pin goes to a high level. The charge controller N-MOSFET, Q2, is turned on again.

### Excess-current Discharge/Battery Short Protection

#### • Protection condition

FHDW01C supplies two-step excess-current protection.

Under normal conditions, during the battery discharging through a load, the voltage of  $V_M$  pin rises with the discharge current increasing. If the discharge current increases to make the voltage of  $V_M$  pin exceed the excess-current discharge protection threshold ( $V_{EDI}$ ) for more than the excess-current discharge protection delay time ( $t_{EDI}$ ), FHDW01C gets into the excess-current discharge protection. If the discharge current increases continuously to make the voltage of  $V_M$  pin exceed the protection battery short threshold ( $V_{SHORT}$ ), FHDW01C gets into the battery short protection.

When FHDW01C is in the excess-current discharge protection or battery short protection, the output of  $D_{OUT}$  pin changes from a high level to a low level ( $V_{SS}$ ).

The external discharge controller N-MOSFET Q2 is turned off, shutting off the discharge loop.  $V_M$  is connected to the  $V_{SS}$  through the internal resistor  $R_{VMS}$ . Once the discharge load is removed, the level of  $V_M$  pin changes to the level of  $V_{SS}$  pin.

• Release condition

In the excess-current discharge protection or the battery short protection, when the voltage of  $V_M$  pin drops lower than the excess-current discharge protection threshold  $V_{EDI}$  for more than the excess-current discharge release delay time ( $t_{EDIR}$ ), FHDW01C recovers to the normal condition. FHDW01C self-releases under the excess-current discharge protection or the battery short protection when removing all of the discharge loads.

After FHDW01C recovers to the normal condition, the output of  $D_{OUT}$  pin goes to a high level. The charge controller N-MOSFET, Q2, is turned on again.

Charger Detection

When a battery in the over-discharge condition is connected to a charger and provided that the  $V_M$  pin voltage is lower than the charger detect voltage ( $V_{CHG}$ ), the FHDW01C releases the over-discharge condition and turns the discharge controller N-MOSFET, Q1 on when the battery voltage becomes equal to or higher than the over-discharge threshold voltage ( $V_{OD}$ ) since the charger detect function works. This action is called charger detection.

When a battery in the over-discharge condition is connected to a charger and provided that the  $V_M$  pin voltage is not lower than the charger detect voltage ( $V_{CHG}$ ), the FHDW01C releases the over-discharge condition when the battery voltage reaches the over-discharge release threshold voltage ( $V_{ODR}$ ).

0V Battery Charging

• 0V battery charge

This function is used to recharge the battery whose voltage is 0V due to self-discharging. If the battery is charged until  $V_{DD}$  is higher than  $V_M$  about 0V charge threshold ( $V_{0V\_CHG}$ ), the  $C_{OUT}$  pin is connected to the  $V_{DD}$ . If the voltage of the  $C_{OUT}$  pin is high enough to turn on the charge controller N-MOS, Q1, a charging circuit is formed through the diode built in the discharge controller N-MOS, Q2. The battery voltage rises. When  $V_{DD}$  is higher than over-voltage discharge protection threshold ( $V_{OD}$ ), FHDW01C enters the normal condition. The output of discharge control pin ( $D_{OUT}$ ) is high. The discharge controller N-MOS is turned on.

• 0V battery charge inhibition

If 0V battery charge is inhibited, the charge control pin ( $C_{OUT}$ ) is connected to the  $V_M$  pin, when  $V_{DD}$  is lower than the 0V charge inhibition threshold ( $V_{NOCHG}$ ). The charge controller N-MOS is turned off.

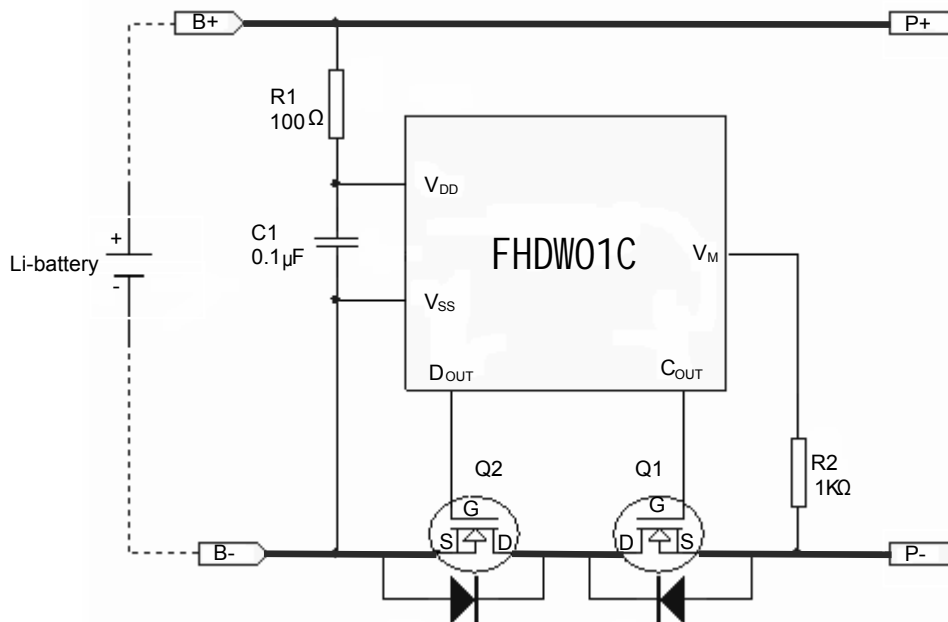


Figure 3 FHDW01C Typical Application Schematic

STATE CONVERSION DIAGRAM

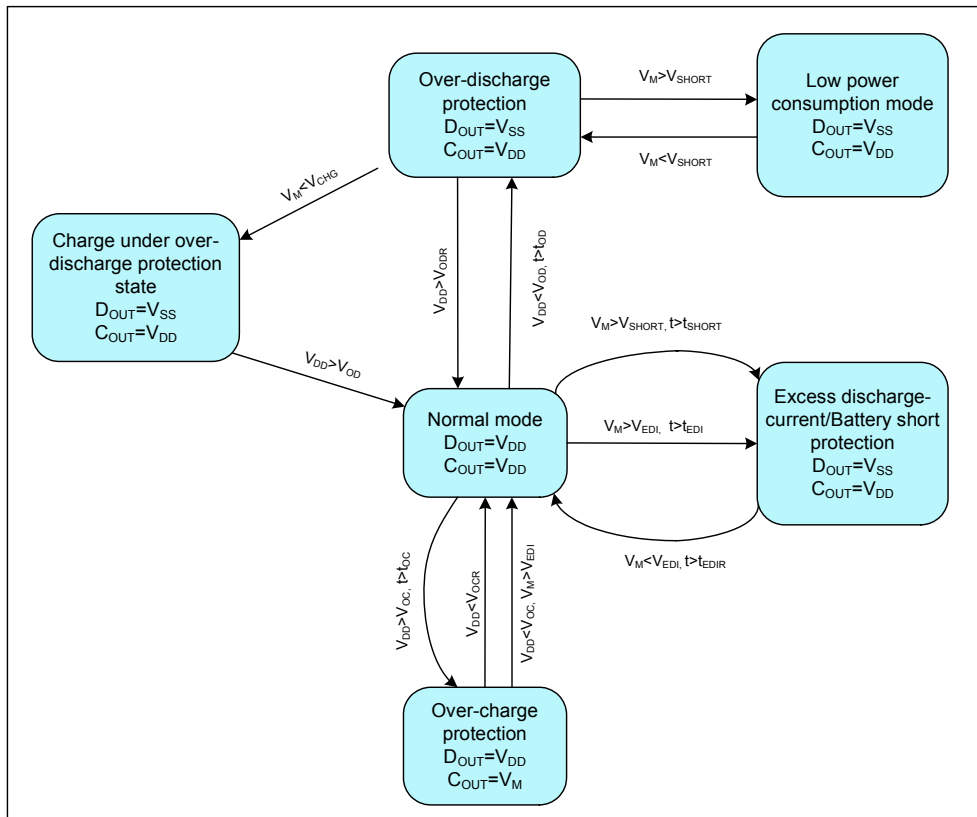


Figure 4 FHDW01C State Conversion Diagram

### STATE CONVERSION AND TIMING DIAGRAM

#### Over-charge/Over-discharge Protection

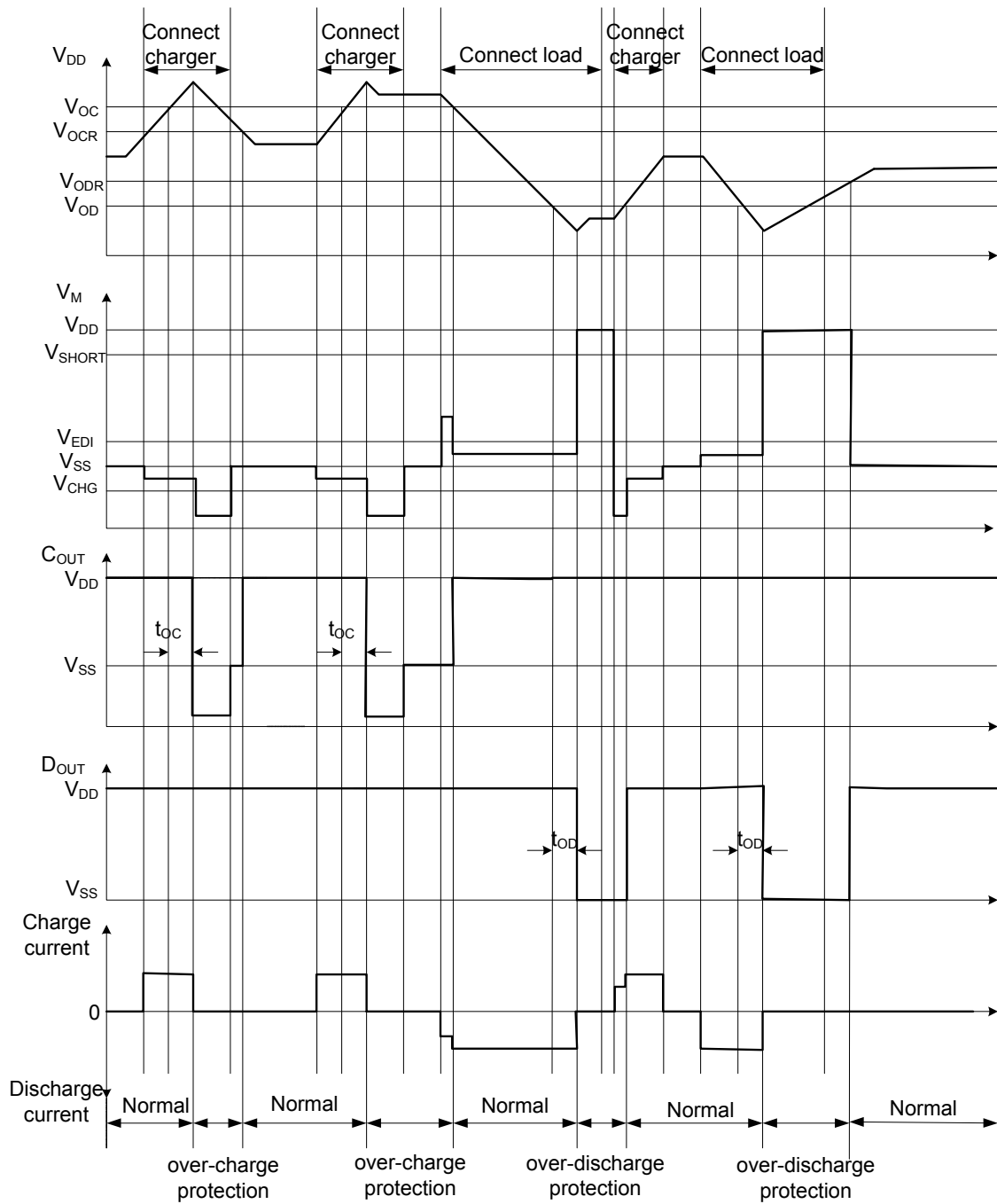


Figure 5 Timing Diagram of Over-charge/Over-discharge Protection



Excess-current Discharge/Battery Short Protection

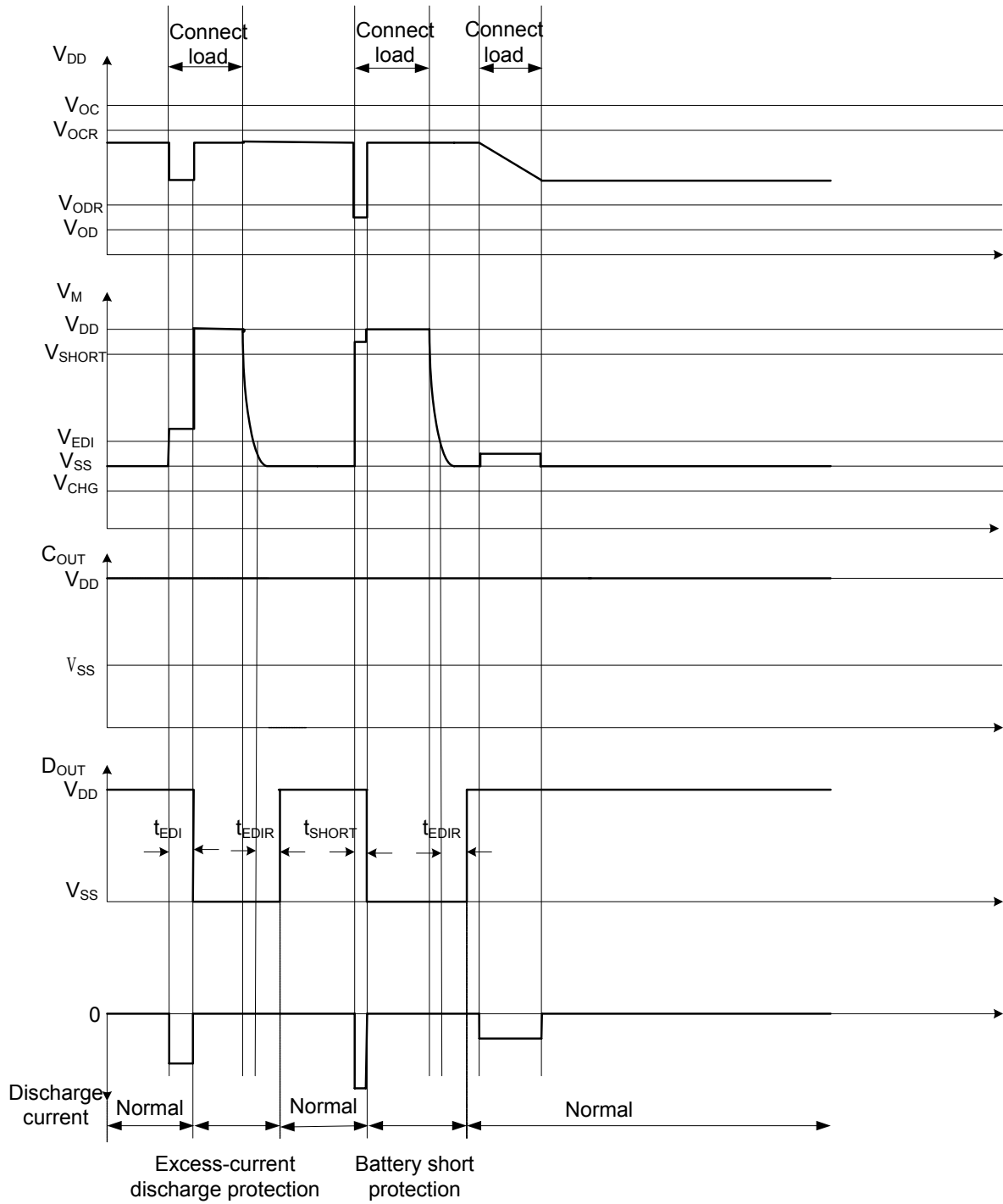


Figure 6 Timing Diagram of Excess-current Discharge/Battery Short Protection

## APPLICATION NOTES

### Selection of Q1 and Q2

Same type of N-MOSFET can be chosen for Q1 and Q2. The threshold voltage,  $V_{th}$  should be between 0.4V and the over-discharge protection threshold voltage ( $V_{OD}$ ). If  $V_{th}$  is less than 0.4V, Q1 might not be turned off. If  $V_{th}$  is higher than  $V_{OD}$ , Q2 might be turned off before the over-discharge is detected.

The breakdown voltages between the gate and the source ( $BV_{GS}$ ) of Q1 and Q2 should be higher than the charger voltage,  $V_{DD}$ . Otherwise, Q1 and Q2 can be destroyed during charging.

### Selection of R1 and R2

The recommended values of R1 and R2 are 100Ω and 1KΩ, respectively. Here R1 should be smaller than R2.

In this application, all the detection thresholds are referred to as  $V_{DD}$ . The  $V_{DD}$  is connected to the battery through a resistor, R1. The difference between each

threshold and the battery voltage increases with the increase of R1. If the charger is connected reversely, the voltage of  $V_{DD}$  pin and  $V_{SS}$  pin may exceed the maximum rating. The IC could be destroyed. Therefore, the value of R1 should be chosen within 500Ω.

If the value of R2 is too small, the current of the IC may be higher than the maximum rating. The IC could be destroyed if the charge is connected reversely. If the value of R2 is too large and a high-voltage charger is connected, the charging current may not be shut off. So the proper value of R2 should be between 500Ω and 2.2KΩ.

### Selection of C1

A filter network is composed of C1 and R1, which filters the power supply. A 0.1μF~1μF ceramic capacitor can be chosen for C1.

## PACKAGE DIMENSION : SOT23-6

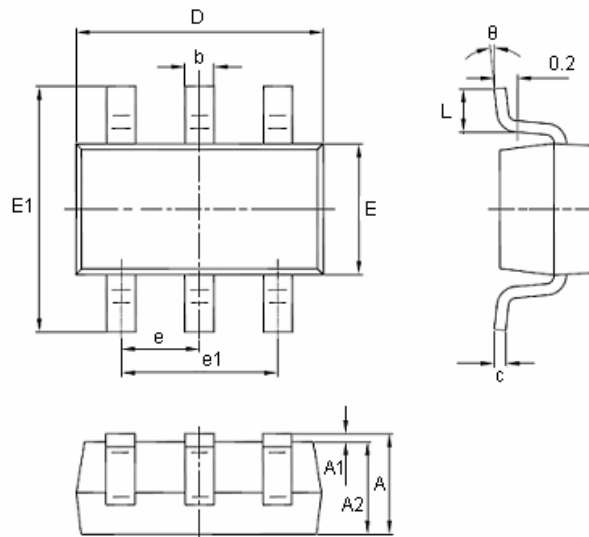


Figure 7 SOT23-6 Package

[Table 5] Physical Dimensions in figure 9 (Unit:mm)

SYMBOL	MIN	MAX
A	1.050	1.250
A1	0.000	0.100
A2	1.050	1.150
b	0.300	0.500
c	0.100	0.200
D	2.280	3.020
E	1.500	1.700
E1	2.650	2.950
e	0.950 (BSC)	
e1	1.800	2.000
L	0.300	0.600
θ	0°	8°